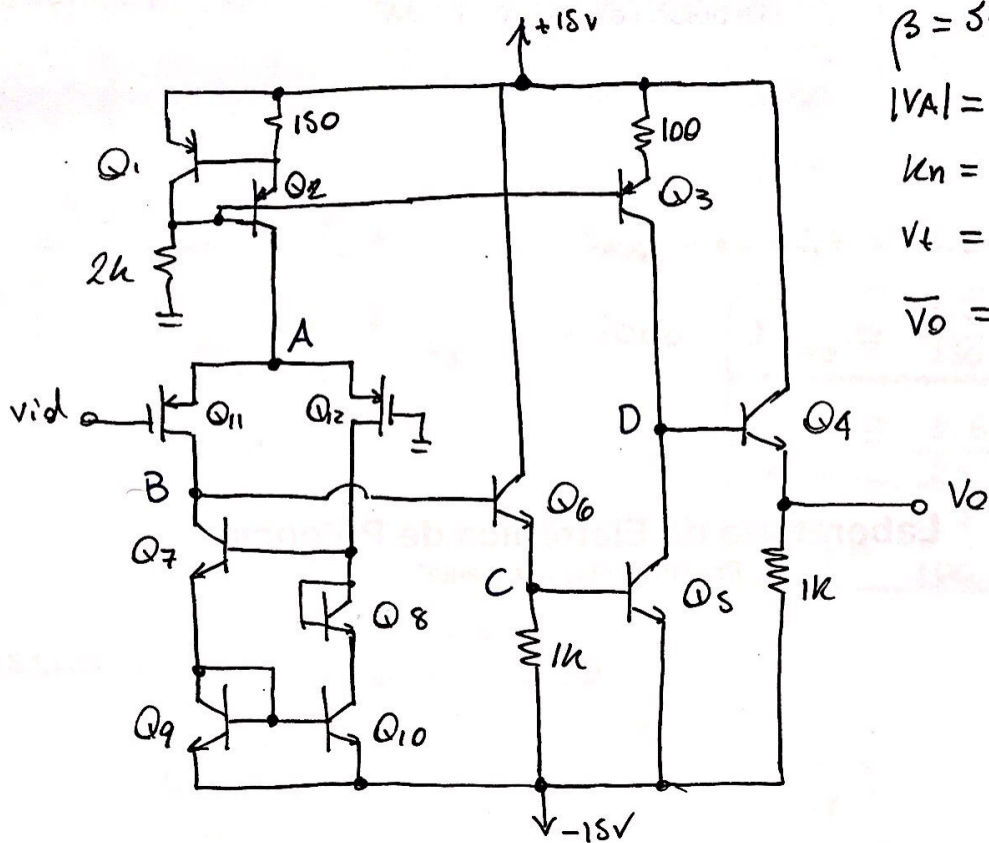


→ Resolução Exercício 6 - CEA

I



$\beta = 500$
 $|V_A| = 50V$
 $k_n = 2mA/V^2$
 $V_t = 1V$
 $\bar{V}_0 = 0V$

a) Análise C.C. → calcular a corrente em todos os transistores e as tensões, A, B, C e D.

→ Como $V_0 = 0$, têm-se que

• $I_{EQ4} = \frac{V_0 - (-15V)}{1k\Omega} = 15mA \rightarrow \text{como } \beta \gg 1 \rightarrow \boxed{I_{CQ4} \cong 15mA}$

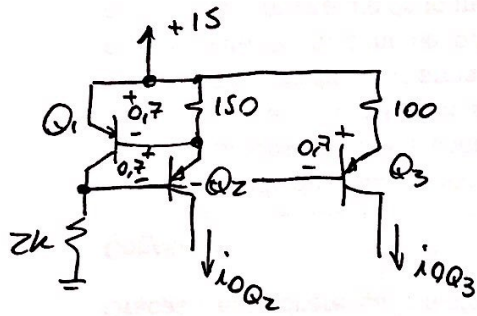
• $V_D = V_0 + 0,7 \rightarrow \boxed{V_D = 0,7V}$

• $V_C = -15V + 0,7 \rightarrow \boxed{V_C = -14,3V}$

$\boxed{I_{EQ6} \cong I_{CQ6} = \frac{V_C - (-15)}{1k} = 0,7mA}$

• $V_B = V_C + 0,7 \Rightarrow \boxed{V_B = -13,6V}$

→ Para encontrar V_A e as demais correntes, deve-se analisar as fontes de corrente:



$$V_{R150} = 0,7V$$

$$V_{R100} = 0,7V$$

$$V_{R2k} = 15 - 1,4 = 13,6V$$

→ logo

$$I_{CQ2} \cong \frac{0,7V}{150} = 4,67mA$$

$$I_{CQ1} \cong \frac{13,6}{2k} = 6,8mA$$

$$I_{CQ3} \cong \frac{0,7}{100} = 7,0mA$$

→ Assim $I_{CQ5} = I_{CQ3} = 7,0mA$

$$I_{DQ11} = I_{DQ12} = \frac{I_{CQ2}}{2} = 2,335mA$$

$$I_{CQ7} = I_{CQ9} = I_{DQ11} = 2,335mA$$

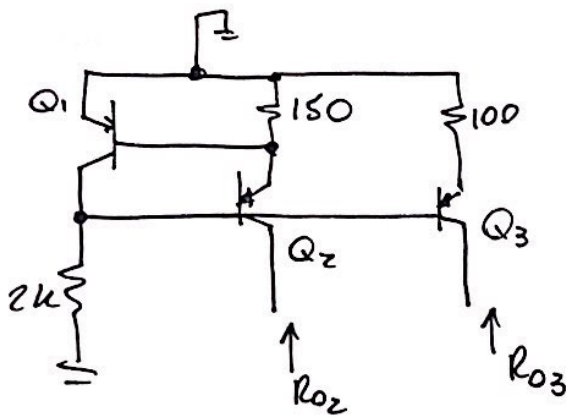
$$I_{CQ8} = I_{CQ10} = I_{DQ12} = 2,335mA$$

→ olhando $Q_{11} \rightarrow I_D = \frac{1}{2} k_n V_{ov}^2 \rightarrow V_{ov} = \sqrt{\frac{2I_D}{k_n}} = 1,528V$

$$V_{SG_{11}} = \sqrt{\frac{2I_{DQ11}}{k_n}} + V_{th} \Rightarrow V_{SG_{11}} = 2,52V$$

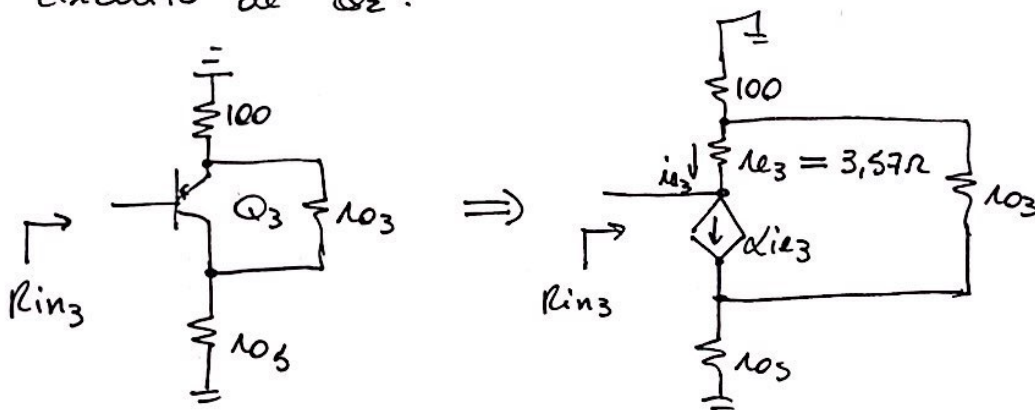
→ logo = $V_A = 2,52V$

b) Cálculo das resistências de saída das fontes de corrente:



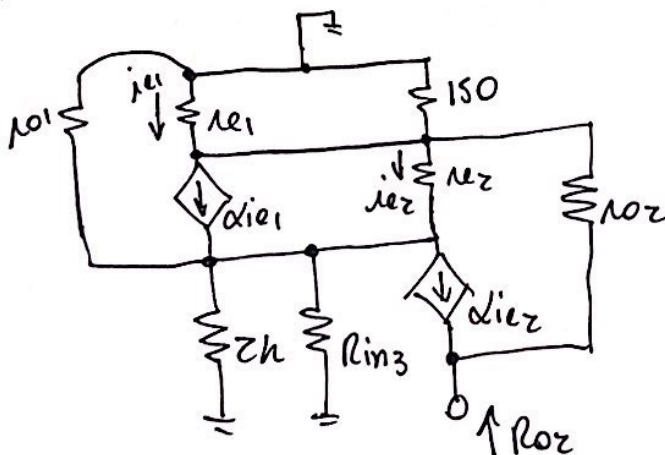
→ Cálculo de R_{02}

→ Primeiramente, levantamos a influência de Q_3 no circuito de Q_2 :



→ Sabemos que $R_{in3} \cong (\beta + 1)(r_{e3} + 100) \cong 52k\Omega$

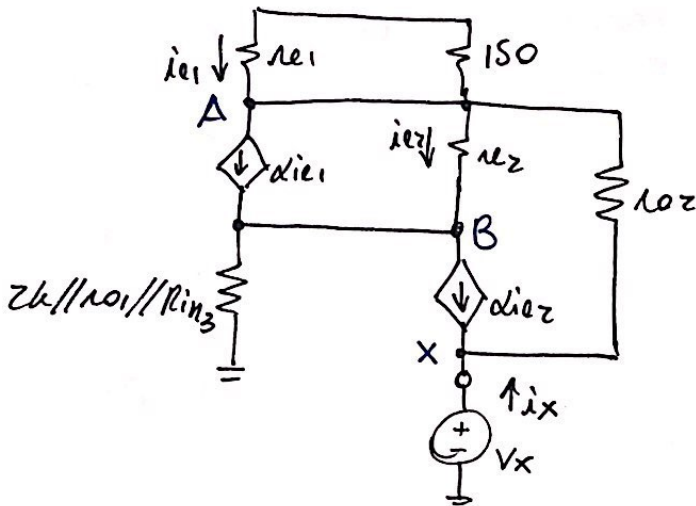
→ Agora levantamos o modelo de pequenos sinais do circuito da fonte:



onde:

$R_{in3} = 52k\Omega$
$r_{e1} = 3,68\Omega$
$r_{e2} = 5,35\Omega$
$r_{o1} = 7,35k\Omega$
$r_{o2} = 10,7k\Omega$
$\alpha = 0,998$

→ Redesenhando



→ $\underline{2k // 10k_1 // R_{in3} = 1,53k\Omega}$

• $i_{e1} = -\frac{V_A}{r_{e1}}$

• $i_{e2} = \frac{V_A - V_B}{r_{e2}}$

→ Resolvendo por método dos nós:

Nó A → $\frac{V_A}{150} + \frac{V_A - V_B}{r_{e2}} + \frac{V_A - V_X}{r_{o2}} + \alpha \left(-\frac{V_A}{r_{e1}} \right) = \left(-\frac{V_A}{r_{e1}} \right)$

$$V_A \left(\frac{1}{150} + \frac{1}{r_{e2}} + \frac{1}{r_{o2}} + \frac{1-\alpha}{r_{e1}} \right) = \frac{V_B}{r_{e2}} + \frac{V_X}{r_{o2}}$$

→ $1-\alpha = 1 - \frac{\beta}{\beta+1} = \frac{1}{\beta+1}$ e $\underline{\underline{(\beta+1)r_{e1} = 1\pi_1}}$

logo, • $V_A \left(\underbrace{\frac{1}{150} + \frac{1}{r_{e2}} + \frac{1}{r_{o2}} + \frac{1}{1\pi_1}}_{\sim \frac{1}{r_{e2}} \text{ (termo dominante)}} \right) = \frac{V_B}{r_{e2}} + \frac{V_X}{r_{o2}}$

$V_A \cong V_B + \frac{V_X r_{e2}}{r_{o2}}$

Nó B → $\frac{V_A - V_B}{r_{e2}} + \alpha \left(-\frac{V_A}{r_{e1}} \right) = \alpha \left(\frac{V_A - V_B}{r_{e2}} \right) + \frac{V_B}{1,53k}$

$$V_A \left(\frac{1}{r_{e2}} - \frac{\alpha}{r_{e1}} - \frac{\alpha}{r_{e2}} \right) = V_B \left(\frac{1}{1,53k} - \frac{\alpha}{r_{e2}} + \frac{1}{r_{e2}} \right)$$

$$V_A \left(\frac{1}{1\pi_2} - \frac{\alpha}{r_{e1}} \right) = V_B \left(\frac{1}{1,53k} + \frac{1}{1\pi_2} \right)$$

$$(V_B + \frac{r_{e2}}{r_{o2}} V_x) \left(\frac{1}{r_{\pi2}} - \frac{\alpha}{r_{e1}} \right) = V_B \left(\frac{1}{1.53k} + \frac{1}{r_{\pi2}} \right)$$

$$V_B \left(\frac{1}{1.53k} + \frac{1}{r_{\pi2}} - \frac{1}{r_{\pi2}} + \frac{\alpha}{r_{e1}} \right) = \left[\frac{r_{e2}}{r_{o2} r_{\pi2}} - \frac{\alpha r_{e2}}{r_{o2} r_{e1}} \right] V_x$$

$$V_B \frac{r_{e1} + \alpha 1.53k}{1.53k r_{e1}} = \frac{r_{e1} r_{e2} - \alpha r_{e2} r_{\pi2}}{r_{o2} r_{\pi2} r_{e1}} V_x$$

$$V_B = \frac{1.53k [r_{e1} r_{e2} - \alpha (\beta + 1) r_{e2}^2]}{r_{o2} r_{e2} (\beta + 1) (r_{e1} + \alpha 1.53k)} V_x$$

$$V_B = \frac{1.53k (r_{e1} - \beta r_{e2})}{r_{o2} (r_{\pi1} + \beta 1.53k)} V_x$$

$$V_A = \left[\frac{1.53k (r_{e1} - \beta r_{e2})}{r_{o2} (r_{\pi1} + \beta 1.53k)} + \frac{r_{e2}}{r_{o2}} \right] V_x$$

$$V_A = \frac{1.53k r_{e1} - (\beta 1.53k r_{e2} + r_{e2} r_{\pi1} + \beta 1.53k r_{e2})}{r_{o2} (r_{\pi1} + \beta 1.53k)} \cdot V_x$$

$$V_A = \frac{r_{e1} (1.53k + r_{\pi2})}{r_{o2} (r_{\pi1} + \beta 1.53k)} V_x$$

• No^x $i_x + \alpha \left(\frac{V_A - V_B}{r_{e2}} \right) + \frac{V_A - V_x}{r_{o2}} = 0$

$$i_x + V_A \left(\frac{\alpha}{r_{e2}} + \frac{1}{r_{o2}} \right) - \frac{\alpha}{r_{e2}} V_B = \frac{V_x}{r_{o2}}$$

$$i_x + \left(\frac{\alpha r_{o2} + r_{e2}}{r_{e2} r_{o2}} \right) \frac{r_{e1} (1.53k + r_{\pi2})}{r_{o2} (r_{\pi1} + \beta 1.53k)} V_x - \frac{\alpha 1.53k (r_{e1} - \beta r_{e2})}{r_{o2} r_{e2} (r_{\pi1} + \beta 1.53k)} V_x = \frac{V_x}{r_{o2}}$$

$$\frac{i_x}{V_x} = \left[\frac{1}{R_{O2}} + \frac{\alpha \cdot 1,53k (\lambda e_1 - \beta \lambda e_2)}{R_{O2} R_{E2} (\lambda \pi_1 + \beta \cdot 1,53k)} - \frac{\lambda e_1 (\alpha R_{O2} + R_{E2}) (1,53k + \lambda \pi_2)}{R_{O2}^2 R_{E2} (\lambda \pi_1 + \beta \cdot 1,53k)} \right]$$

$$= \frac{R_{O2} R_{E2} (\lambda \pi_1 + \beta \cdot 1,53k) + \alpha \cdot 1,53k (\lambda e_1 - \beta \lambda e_2) R_{O2} - \lambda e_1 (\alpha R_{O2} + R_{E2}) (1,53k + \lambda \pi_2)}{R_{O2}^2 R_{E2} (\lambda \pi_1 + \beta \cdot 1,53k)}$$

$$\frac{i_x}{V_x} = \frac{1,53k (\beta R_{O2} R_{E2} + \alpha \cancel{R_{O2} \lambda e_1} - \alpha \beta R_{O2} R_{E2} - \alpha \cancel{R_{O2} \lambda e_1} - \lambda e_1 R_{E2})}{R_{O2}^2 R_{E2} (\lambda \pi_1 + \beta \cdot 1,53k)} +$$

$$\frac{R_{O2} R_{E2} \lambda e_1 (\beta + \lambda) - \alpha (\beta + \lambda) R_{O2} \lambda e_1 R_{E2} - (\beta + \lambda) \lambda e_1 R_{E2}^2}{R_{O2}^2 R_{E2} (\lambda \pi_1 + \beta \cdot 1,53k)}$$

• Manipulando esta expressão, encontramos:

$$\frac{i_x}{V_x} = \frac{\cancel{R_{O2} R_{E2}} [1,53k (\alpha - \lambda e_1 / R_{O2}) + \lambda e_1 (1 - \lambda \pi_2 / R_{O2})]}{R_{O2}^2 R_{E2} (\lambda \pi_1 + \beta \cdot 1,53k)}$$

$$R_{O2} = \frac{V_x}{i_x} = \frac{R_{O2} (\lambda \pi_1 + \beta \cdot 1,53k)}{1,53k (\alpha - \lambda e_1 / R_{O2}) + \lambda e_1 (1 - \lambda \pi_2 / R_{O2})} = \boxed{5,355 \text{ M}\Omega}$$

• mas, $\alpha - \frac{\lambda e_1}{R_{O2}} = \frac{\alpha R_{O2} - \lambda e_1}{R_{O2}} \approx \alpha$, pois $R_{O2} \gg \lambda e_1$

$$\text{logo} \rightarrow R_{O2} \approx \frac{R_{O2} (\lambda \pi_1 + \beta \cdot 1,53k)}{\alpha \cdot 1,53k + \lambda e_1 (1 - \lambda \pi_2 / R_{O2})}$$

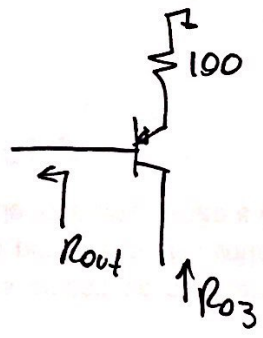
\rightarrow novamente $\rightarrow \alpha \cdot 1,53k \gg \lambda e_1 (1 - \lambda \pi_2 / R_{O2})$, pois

$$\frac{\lambda \pi_2}{R_{O2}} < 1 \text{ e } 1,53k \Omega \gg \lambda e_1$$

assim:

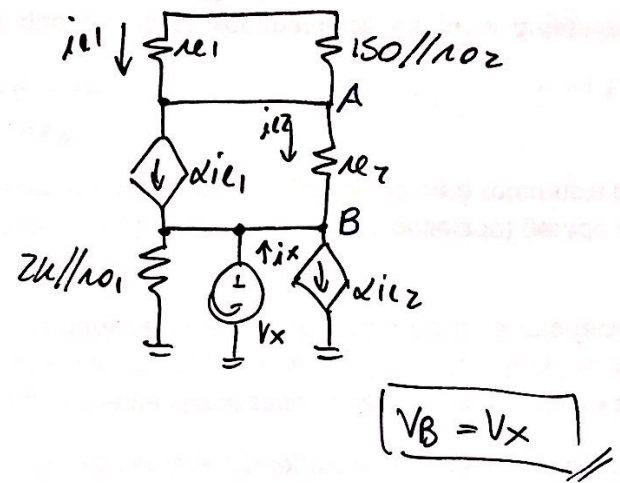
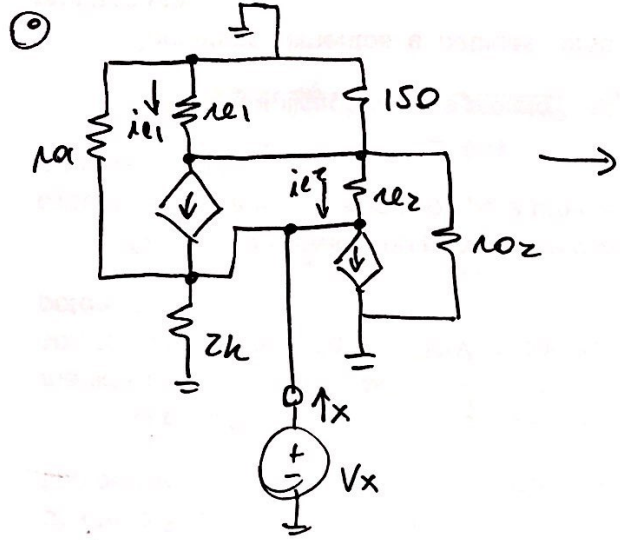
$$\boxed{R_{O2} \approx \frac{R_{O2} (\lambda \pi_1 + \beta \cdot 1,53k)}{\alpha \cdot 1,53k}} = 5,37 \text{ M}\Omega$$

• Cálculo de R_{o3}



→ Qual a influência de Q_2 em Q_3 ? → calcular R_{out}

Calculando R_{out}



→ No' A
$$\frac{V_A}{150 // 100} + \frac{V_A - V_B}{100} + \alpha \left(-\frac{V_A}{100} \right) = \left(-\frac{V_A}{100} \right)$$

$$V_A \left(\frac{1}{150 // 100} + \frac{1}{100} + \frac{1}{100} \right) = \frac{V_B}{100}$$

$\sim \frac{1}{100}$

$V_A \approx V_B$

No' B

$$i_x + \alpha \left(-\frac{V_A}{100} \right) + \frac{V_A - V_B}{100} = \alpha \left(\frac{V_A - V_B}{100} \right) + \frac{V_B}{2k // 100}$$

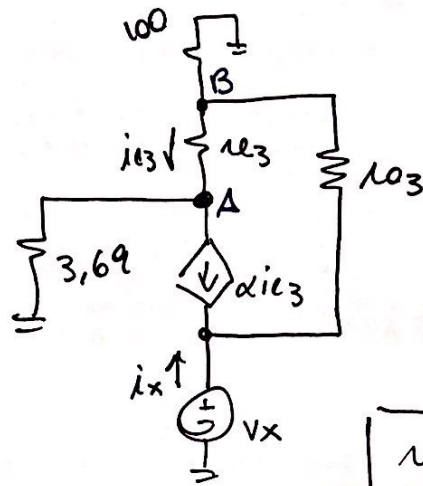
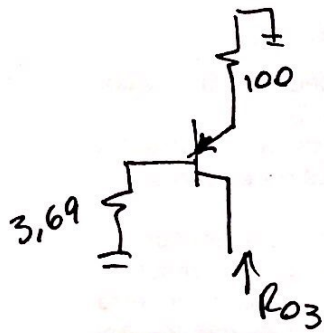
$$i_x = \frac{V_B}{2k // 100} + (\alpha - 1) \left(\frac{V_A - V_B}{100} \right) + \frac{\alpha V_A}{100}$$

$$i_x = V_B \left(\frac{1}{2k // 100} + \frac{\alpha}{100} \right) \Rightarrow R_{out} = \frac{V_B}{i_x} \approx \frac{100}{\alpha} = 3,69 \Omega$$

$\sim \frac{\alpha}{100}$

⇒ Continuando

VIII



$$\begin{aligned} R_3 &= 3,57\Omega \\ R_03 &= 7,14k\Omega \end{aligned}$$

• No A

$$\frac{V_B - V_A}{R_3} = \frac{V_A}{3,69} + \alpha \left(\frac{V_B - V_A}{R_3} \right)$$

$$V_A \left(\frac{1}{3,69} + \frac{1-\alpha}{R_3} \right) = V_B \left(\frac{1-\alpha}{R_3} \right)$$

$$V_A \left(\frac{1}{3,69} + \frac{1}{100} \right) = \frac{V_B}{100}$$

$$V_A \frac{100 + 3,69}{3,69 \cdot 100} = \frac{V_B}{100} \rightarrow \boxed{V_A = \frac{3,69}{100 + 3,69} V_B}$$

$$\boxed{V_A = 2 \times 10^{-3} \cdot V_B}$$

No B:

$$\frac{V_B}{100} + \frac{V_B - V_A}{R_3} + \frac{V_B - V_x}{R_03} = 0$$

$$V_B \left(\frac{1}{100} + \frac{0,998}{R_3} + \frac{1}{R_03} \right) = \frac{V_x}{R_03}$$

$$V_B \frac{R_3 R_03 + 0,998 \cdot 100 R_03 + 100 R_3}{100 R_3 R_03} = \frac{V_x}{R_03}$$

$$V_B = \frac{100 \cdot R_3}{99,8 R_03 + 100 R_3 + R_3 R_03} \cdot V_x$$

IV

No X

$$i_x + \alpha \left(\frac{V_B - V_A}{\mu_3} \right) + \frac{V_B - V_X}{\rho_3} = 0$$

$$i_x + \frac{\alpha \cdot 0,998 V_B}{\mu_3} + \frac{V_B}{\rho_3} = \frac{V_X}{\rho_3}$$

$$i_x = \frac{V_X}{\rho_3} - \left(\frac{1}{\rho_3} + \frac{\alpha \cdot 0,998}{\mu_3} \right) \cdot \left(\frac{100 \cdot \mu_3}{99,8 \rho_3 + 100 \mu_3 + \mu_3 \rho_3} \right) V_X$$

$$i_x = \frac{V_X}{\rho_3} - \frac{\mu_3 + \alpha \cdot 0,998 \rho_3}{\mu_3 \rho_3} \cdot \frac{100 \mu_3}{99,8 \rho_3 + 100 \mu_3 + \mu_3 \rho_3} V_X$$

$$\frac{i_x}{V_X} = \frac{99,8 \rho_3 + 100 \mu_3 + \mu_3 \rho_3 - 100 \mu_3 - \alpha \cdot 0,998 \cdot 100 \rho_3}{\rho_3 (99,8 \rho_3 + 100 \mu_3 + \mu_3 \rho_3)}$$

$$= \frac{99,8 \rho_3 (1 - \alpha) + \mu_3 \rho_3}{\rho_3 (99,8 \rho_3 + 100 \mu_3 + \mu_3 \rho_3)} (\beta + 1)$$

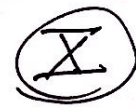
$$\frac{i_x}{V_X} = \frac{99,8 + \mu_3}{(\beta + 1) 99,8 \rho_3 + 100 \mu_3 + \mu_3 \rho_3}$$

$$\rho_3 = \frac{(\beta + 1) 99,8 \rho_3 + (100 + \mu_3) \mu_3}{99,8 + \mu_3} = \underline{196,28 \text{ k}\Omega}$$

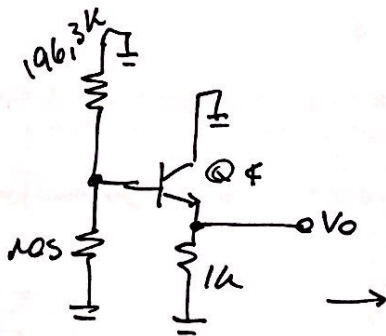
→ Compilando

$$\begin{aligned} \rho_2 &= 5,36 \text{ M}\Omega \\ \rho_3 &= 196,3 \text{ k}\Omega \end{aligned}$$

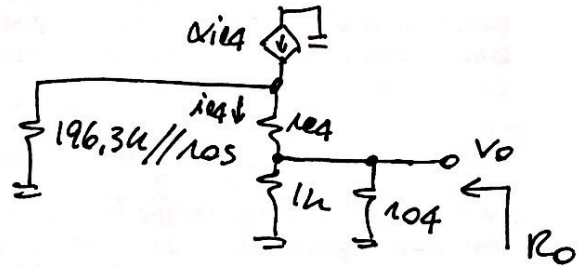
c) Cálculo de A_d e R_o



→ Iniciando com o cálculo de R_o



→ fazendo $v_{id} = 0$, o estágio de saída se torna:



$$\rightarrow R_o = (104k // 1k) // \left(104k + \frac{196.3k // 105k}{\beta + 1} \right)$$

$$\bullet R_{05} = \frac{50}{7mA} = 7.14k\Omega$$

$$\bullet R_{04} = \frac{50}{15mA} = 3.33k\Omega$$

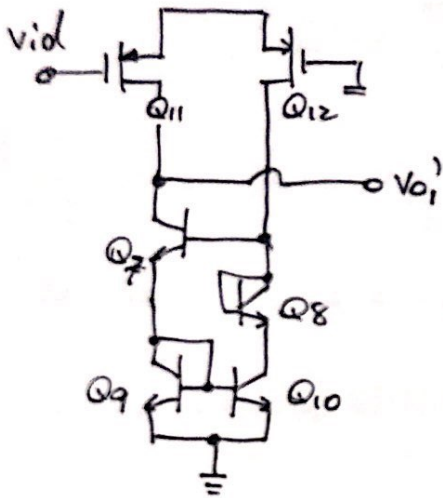
$$\bullet r_{eq} = \frac{28mV}{15mA} = 1.7\Omega$$

$$\boxed{R_o = 15.15\Omega}$$

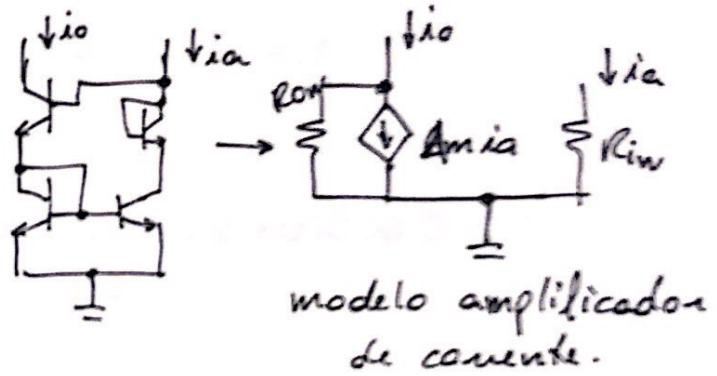
→ Cálculo de A_d

↳ P/ realizar o cálculo do ganho diferencial é interessante analisar os múltiplos estágios do amplificador e em seguida multiplicar os ganhos individuais de cada estágio p/ encontrar o ganho total.

→ Analisando o par diferencial com carga ativa:



→ Uma boa estratégia é fazer um modelo do espelho de Wilson:

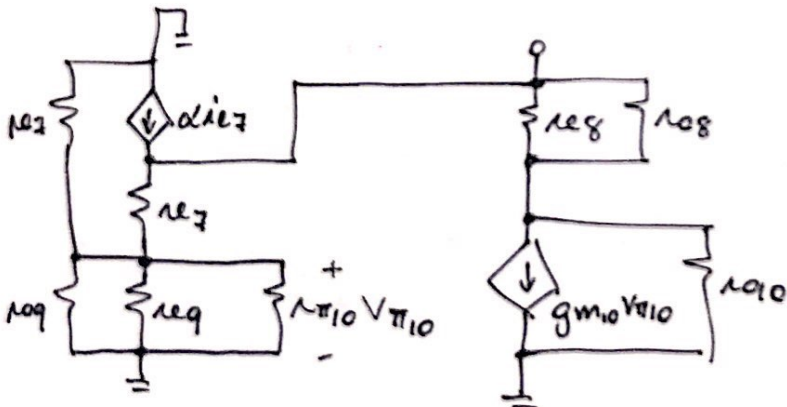


→ Em função das análises feitas em sala, sabe-se que

$$R_{0w} \approx \frac{\beta \cdot r_{07}}{2}; \text{ sendo que nos resta calcular}$$

R_{iw} e A_m .

→ cálculo de R_{iw}



→ $r_{07} = r_{08} = r_{09} = r_{010} = r_0 = \frac{50V}{2.335mA}$

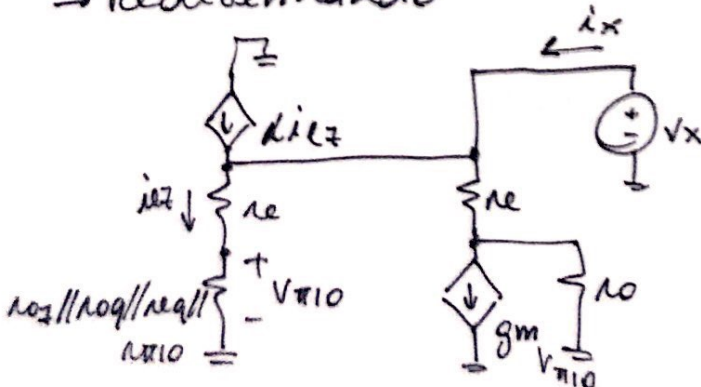
• $r_0 = 21,4k\Omega$

• $r_{e7} = r_{e10} = r_e = \frac{25mV}{2.335mA} = 10,7\Omega$

→ $r_e // r_0 \approx r_e$

• $g_{m10} = \frac{2.335mA}{25mV} = 0,093A/V$

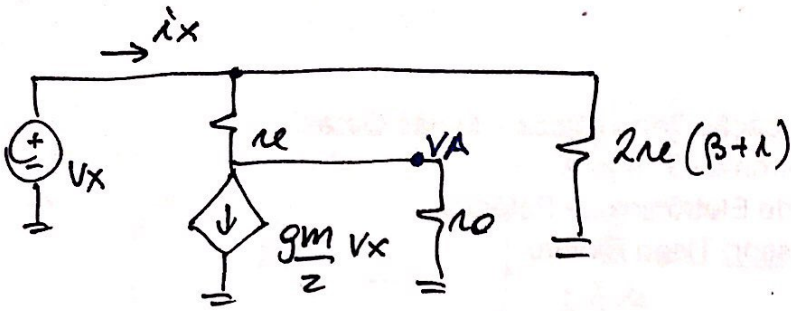
→ Redesenhando



→ $r_{07} // r_{09} // r_{e9} // r_{\pi10} \approx r_e$

• Assim:

$$V_{\pi10} \approx \frac{v_x \cdot r_e}{2r_e} \approx \frac{v_x}{2}$$



No A

$$\frac{V_x - V_A}{r_e} = \frac{V_A}{r_o} + \frac{g_m V_x}{z}$$

$$V_x - V_A = \frac{r_e}{r_o} V_A + \frac{g_m r_e}{z} V_x$$

$$\rightarrow g_m = \frac{\alpha}{r_e}$$

$$V_x - V_A = \frac{r_e}{r_o} V_A + \frac{\alpha}{z} V_x$$

$$V_x \left(1 - \frac{\alpha}{z}\right) = V_A \frac{r_o + r_e}{r_o}$$

$$\boxed{V_A = \frac{r_o \left(1 - \frac{\alpha}{z}\right)}{r_o + r_e} V_x}$$

$$\rightarrow \alpha \approx 1$$

$$\Rightarrow$$

$$\boxed{V_A \approx \frac{V_x}{z}}$$

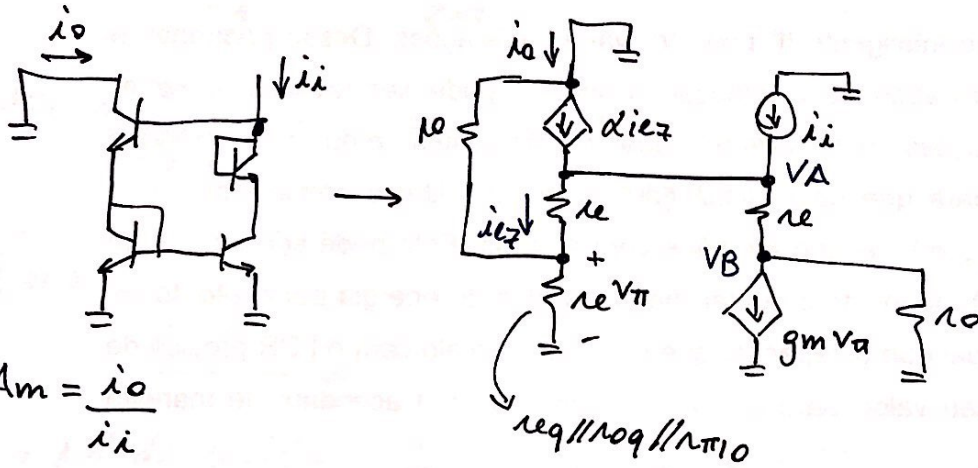
$$\text{No } V_x \rightarrow i_x = \frac{V_x - V_A}{r_e} + \frac{V_A}{2r_e(\beta+1)}$$

$$i_x = \frac{1}{2} r_e \cdot V_x + \frac{V_x}{2r_e(\beta+1)}$$

$$\frac{i_x}{V_x} = \frac{\beta+1 + 1}{2r_e(\beta+1)} = \frac{\beta+2}{2r_e(\beta+1)}$$

$$\boxed{R_{iw} = \frac{2r_e(\beta+1)}{\beta+2} = 21.36 \Omega}$$

→ Cálculo de Am



$$A_m = \frac{i_o}{i_i}$$

$$\rightarrow V_{\pi} = V_A \cdot \frac{r_e // r_o}{r_e + r_e // r_o} \approx \frac{V_A}{2}$$

Nó A

$$i_i = \frac{V_A - V_{\pi}}{r_e} + \frac{V_A - V_B}{r_e} - \alpha \left(\frac{V_A - V_{\pi}}{r_e} \right)$$

$$i_i = \frac{V_A(1-\alpha)}{2r_e} + \frac{V_A}{r_e} - \frac{V_B}{r_e} = \frac{3V_A}{2r_e} - \frac{V_B}{r_e}$$

$$i_i = V_A \left(\frac{1}{2r_{\pi}} + \frac{1}{r_e} \right) - \frac{V_B}{r_e}$$

Nó B

$$\frac{V_A - V_B}{r_e} = \frac{V_B}{r_o} + \frac{g_m V_A}{2}$$

$$V_A - V_B = \frac{r_e}{r_o} V_B + \frac{g_m r_e}{2} V_A$$

$$V_A \left(1 - \frac{g_m r_e}{2} \right) = V_B \left(1 + \frac{r_e}{r_o} \right)$$

$$V_A \left(1 - \frac{\alpha}{2} \right) = V_B \left(\frac{r_o + r_e}{r_o} \right)$$

aproximando
 $\alpha \approx 1$
 $r_o \gg r_e$

$$\rightarrow \frac{V_A}{2} = V_B \rightarrow \boxed{V_B = \frac{V_A}{2}}$$

logo $\rightarrow i_i = V_A \frac{1 + 2(\beta+1)}{2r_{\pi}} - \frac{V_A}{2r_e}$

$$i_i = V_A \frac{\cancel{1} + 2\beta + 2 - \beta}{2r_{\pi}} \rightarrow \boxed{V_A = \frac{2r_{\pi}}{\beta+2} i_i}$$

$$i_{e7} = \frac{V_A - V_{\pi}}{r_e} = \frac{V_A}{2r_e}$$

$$i_{e7} = \frac{2r_{\pi}}{2r_e(\beta+2)} \cdot i_i = \frac{2r_e(\beta+1)}{2r_e(\beta+2)} i_i$$

$$i_{e7} = \frac{\beta+1}{\beta+2} i_i$$

→ Nó de saída $i_o = \alpha i_{e7} + \frac{0 - V_{\pi}}{r_o}$

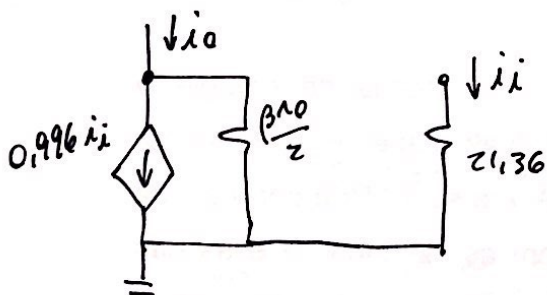
$$i_o = \frac{\beta}{\beta+1} \cdot \frac{\beta+1}{\beta+2} i_i - \frac{V_A}{2r_o}$$

$$i_o = \frac{\beta}{\beta+2} i_i - \frac{2r_{\pi}}{2r_o(\beta+2)} i_i$$

$$\frac{i_o}{i_i} = \frac{\beta r_o - r_{\pi}}{r_o(\beta+2)} \rightarrow \text{como } \beta r_o \gg r_{\pi}$$

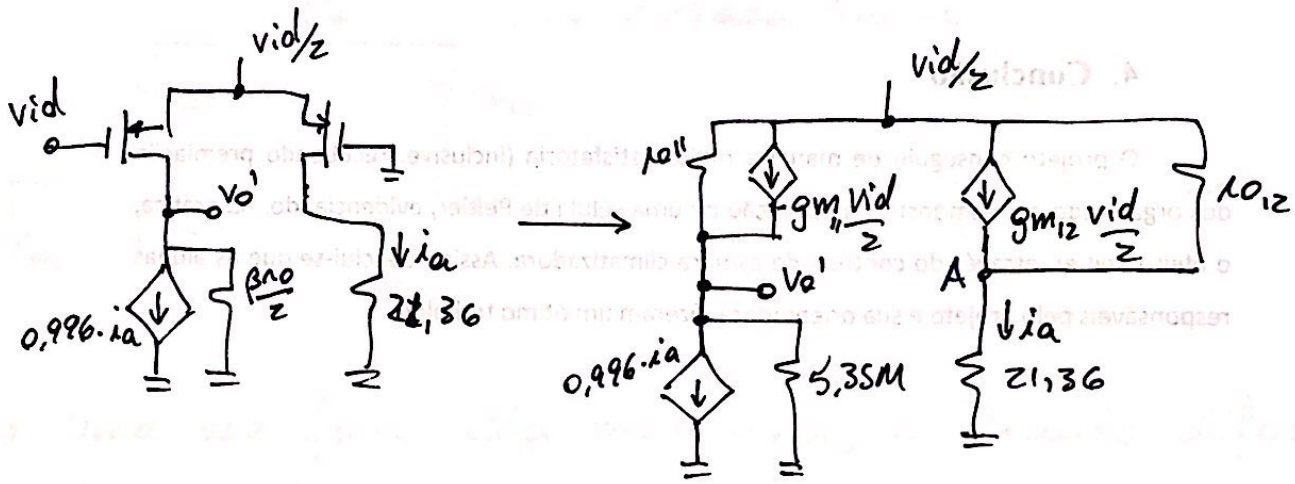
$$A_m = \frac{i_o}{i_i} \approx \frac{\beta}{\beta+2} = 0,996$$

→ Assim, o modelo do espelho de Wilson se torna



$$\beta \frac{r_o}{2} = 5,35 M\Omega$$

→ Uma vez definido o modelo de espelho, voltamos ao circuito do par diferencial



- No A:

$$g_{m12} \frac{v_{id}}{2} + \frac{v_{id} - V_A}{r_{o12}} = \frac{V_A}{21,36}$$

$$\left(g_{m12} + \frac{1}{r_{o12}} \right) \frac{v_{id}}{2} = V_A \left(\frac{1}{21,36} + \frac{1}{r_{o12}} \right)$$

$\sim \frac{1}{21,36}$

- $g_{m11} = g_{m12} = \frac{2I_{D11}}{V_{ov}} = 3,06 \text{ mA/V}$
- $r_{o11} = r_{o12} = \frac{50}{I_{D11}} = 21,4 \text{ k}\Omega$

$$V_A = \frac{21,36}{2} \cdot \left(\frac{1 + g_{m12} r_{o12}}{r_{o12}} \right) v_{id}$$

$$i_a = \frac{V_A}{21,36} \rightarrow i_a = \frac{1 + g_{m12} r_{o12}}{2 r_{o12}} v_{id}$$

- No vo':

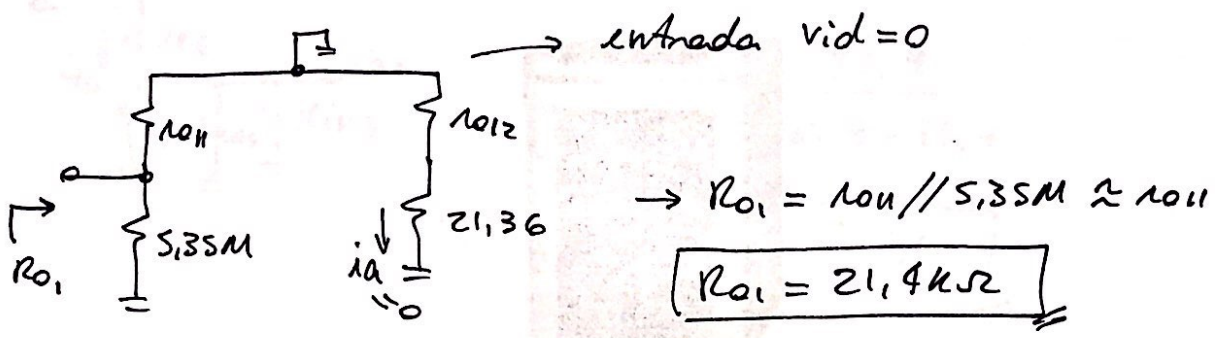
$$0,996 i_a + \frac{v_{o'}}{5,35 \text{ M}} + \frac{v_{o'} - v_{id}/2}{r_{o11}} = -g_{m11} \frac{v_{id}}{2}$$

$$v_{o'} \left(\frac{1}{5,35 \text{ M}} + \frac{1}{r_{o11}} \right) = v_{id} \left(-\frac{g_{m11}}{2} + \frac{1}{2 r_{o11}} - \frac{0,996(1 + g_{m12} r_{o12})}{2 r_{o12}} \right)$$

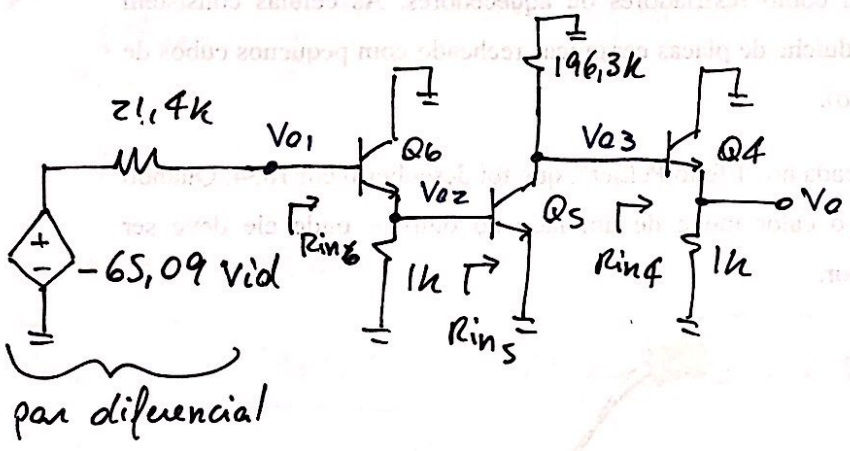
$$v_{o'} = \left(\frac{5,35 \text{ M} \cdot r_{o11}}{5,35 \text{ M} + r_{o11}} \cdot \frac{-g_{m11} r_{o11} + 1 - 0,996 - g_{m12} r_{o12} \cdot 0,996}{2 r_{o11}} \right) v_{id}$$

$$\frac{v_{o'}}{v_{id}} = -65,09 \text{ V/V}$$

→ Podemos também calcular a resistência de saída do par:

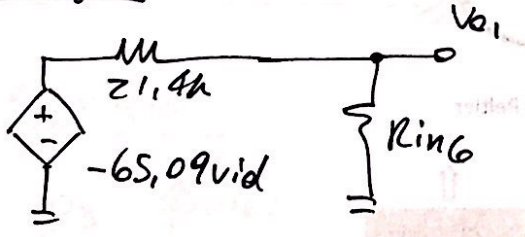


→ Uma vez feita esta modelagem, o circuito diferencial do amplificador se torna:



→ calculando os ganhos:

1º estágio:



$$A_{v1} = \frac{V_{o1}}{V_{id}} = \frac{R_{in6}}{R_{in6} + 21,4k} \cdot -65,09$$

$A_{v1} = -61,11 \text{ V/V}$

$$R_{in6} = (\beta + 1) [1k \parallel R_{in5}]$$

$$R_{in5} = (\beta + 1) r_{e5}$$

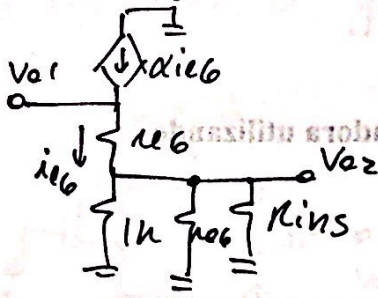
$$r_{e5} = \frac{25mV}{7mA} = 3,57\Omega$$

$$r_{e6} = \frac{25mV}{0,7mA} = 35,7\Omega$$

• logo → $R_{in5} = 1,79k\Omega$

$R_{in6} = 328,7k\Omega$

2º estágio



$$r_{o6} = 71,4 \text{ k}\Omega$$

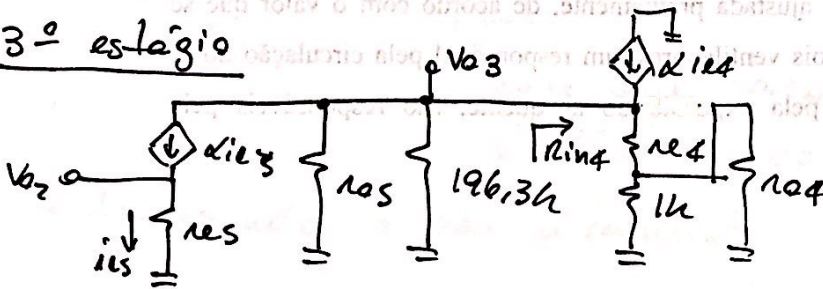
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$$1\text{k} // r_{o6} // R_{ins} = 635,9 \Omega$$

$$V_{02} = V_{01} \cdot \frac{635,9}{635,9 + 35,7}$$

$$A_{V2} = \frac{V_{02}}{V_{01}} = 0,947 \text{ V/V}$$

3º estágio



- $r_{e5} = 3,57 \Omega$
- $r_{o5} = 7,14 \text{ k}\Omega$
- $r_{e4} = 1,7 \Omega$
- $r_{o4} = 3,33 \text{ k}\Omega$

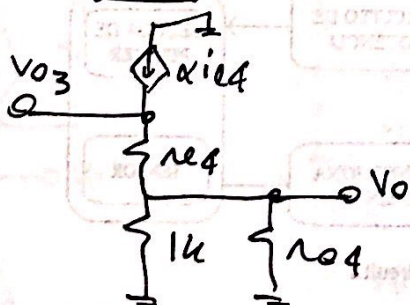
$$R_{in4} = (r_{e4} + 1\text{k} // r_{o4}) (\beta + 1)$$

$$R_{in4} = 386,1 \text{ k}\Omega$$

$$\frac{V_{03}}{V_{02}} = \frac{-\alpha (r_{o5} // 196,3\text{k} // R_{in4})}{r_{e5}} = -1,896 \text{ kV/V}$$

$$A_{V3} = -1,896 \text{ kV/V}$$

4º estágio



$$V_0 = V_{03} \cdot \frac{1\text{k} // r_{o4}}{1\text{k} // r_{o4} + r_{e4}}$$

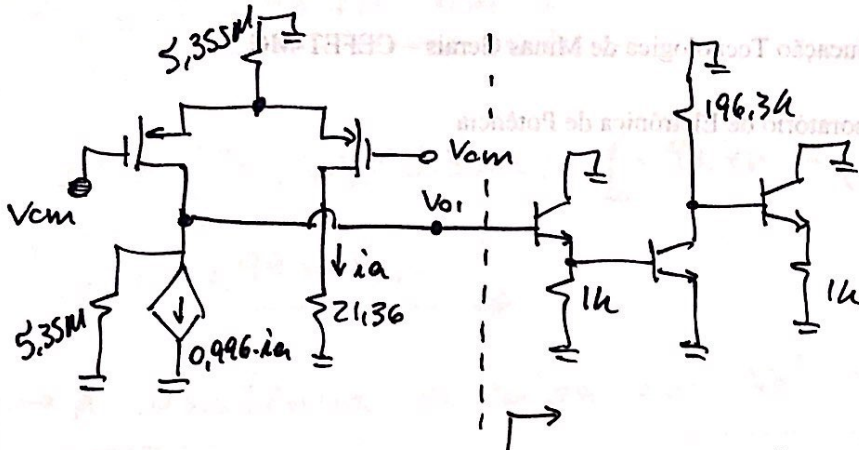
$$A_{V4} = \frac{V_0}{V_{03}} = 0,998 \text{ V/V}$$

• Assim, $A_d = A_{V1} \cdot A_{V2} \cdot A_{V3} \cdot A_{V4}$

$$A_d = 109,5 \text{ kV/V}$$

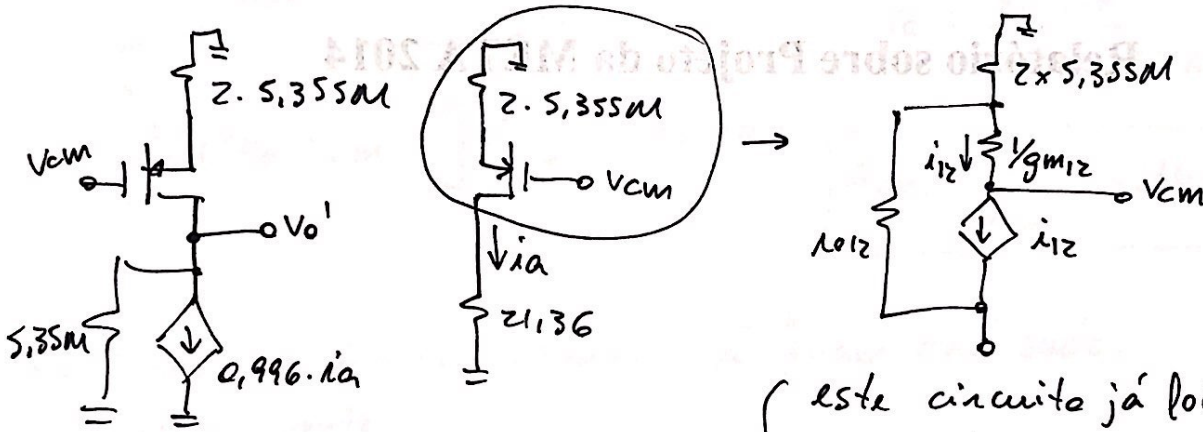
d) Calcular o ganho de modo-comum

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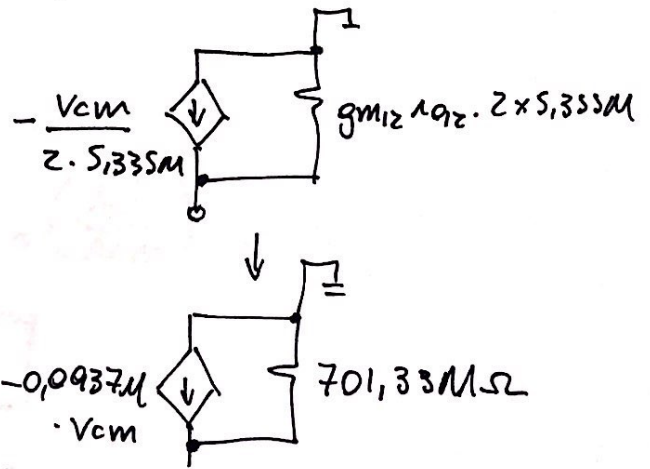
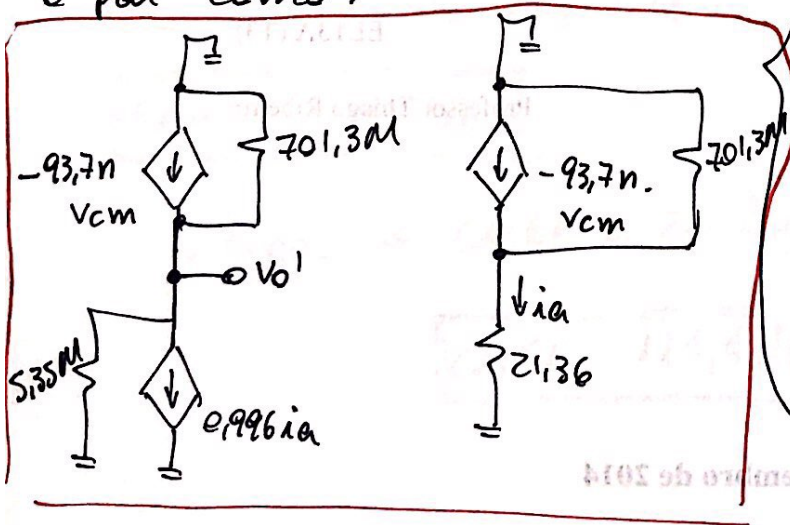
note que daqui p/ frente a análise é idêntica ao modo diferencial.

→ Analisando o par diferencial



este circuito já foi analisado nas aulas. Gerou-se o seguinte modelo:

Assim, pode-se redesenhar o par como:



→ $i_a \cong -93,7 n \cdot V_{cm}$

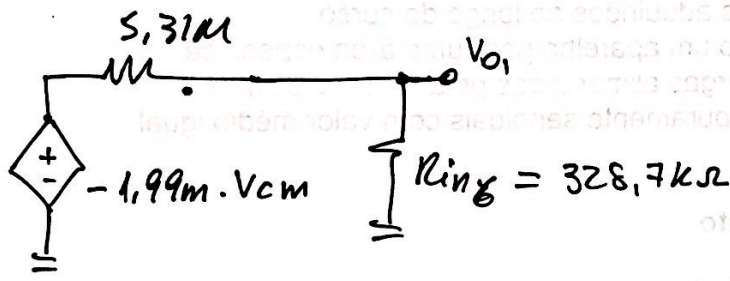
$V_{o1} = (701,3M // 5,35M) \cdot [-93,7 n - (-93,7 n \cdot 0,996)] V_{cm}$

$V_{o1} = -1,99 m V_{cm}$

→ A resistência de saída em V_{o1} se torna

$R_{o1} = 701,3M // 5,35M = 5,31 M\Omega$

→ Assim o ganho do 1º estágio se torna:



$V_{o1} = \frac{328,7k}{328,7k + 5,31M} \cdot -1,99 m V_{cm}$

$A_{cm1} = \frac{V_{o1}}{V_{cm}} = -116 MV/V$

→ Como os demais estágios já foram analisados anteriormente

$A_{cm} = A_{cm1} \cdot A_{v2} \cdot A_{v3} \cdot A_{v4}$

$A_{cm} = 0,208 V/V$

• logo. → $CMRR = 20 \log_{10} \left(\frac{A_d}{A_{cm}} \right)$

$CMRR = 114,4 dB$